Office Automation-Use<br>3-Phase Brushless Motor Driver

## Overview

The LB1820 is a three-phase brushless motor with a digital speed control circuit built in.
The LB1820 is ideally suited for use in office automation applications such as laser beam printers and drum motor drivers.

## Features

- Three-phase brushless motor driver with digital speed control function
- 30 V withstand voltage and 2.5 A output current
- Current limiter built in
- Low-voltage protection circuit built in
- Thermal shutdown circuit built in
- Hall amp with hysteresis
- Start/stop pin built in
- Crystal oscillator and divider built in
- Digital speed control circuit built in
- Lock detector built in


## Package Dimensions

unit: mm
3147C


## Specifications

## Absolute Maximum Ratings at $\mathrm{Ta}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :---: | :---: | :---: | :---: | :---: |
| Maximum supply voltage 1 | VCC |  | 30 | V |
| Maximum supply voltage 2 | $\mathrm{V}_{\mathrm{M}}$ |  | 30 | V |
| Output current | lo | $\mathrm{t} \leq 100 \mathrm{~ms}$ | 2.5 | A |
| Allowable power dissipation 1 | Pd max1 | Independent IC | 3 | W |
| Allowable power dissipation 2 | Pd max2 | With arbitrarily large heat sink | 20 | W |
| Operating temperature | Topr |  | -20 to +80 | ${ }^{\circ} \mathrm{C}$ |
| Storage temperature | Tstg |  | -55 to +150 | ${ }^{\circ} \mathrm{C}$ |

Allowable Operating Ranges at $\mathrm{Ta}=25^{\circ} \mathrm{C}$

| Parameter | Symbol | Conditions | Ratings | Unit |
| :--- | :---: | :---: | :---: | :---: |
| Supply voltage range 1 | $\mathrm{V}_{\mathrm{CC}}$ |  | 9.5 to 28 | V |
| Supply voltage range 2 | $\mathrm{V}_{\mathrm{M}}$ |  | 5 to 28 | V |
| Voltage regulator output current | $\mathrm{IVH}_{\mathrm{VH}}$ |  | 0 to +20 | mA |
| Comparator output current | IOSC |  | 0 to +30 | mA |
| Lock detector output current | ILD |  | 0 to +20 | mA |

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## LB1820

Electrical Characteristics at $\mathrm{Ta}=25^{\circ} \mathrm{C}, \mathrm{V}_{\mathrm{CC}}=\mathrm{V}_{\mathrm{M}}=24 \mathrm{~V}$

| Parameter | Symbol | Conditions | min | typ | max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Supply current 1 | ICC1 |  |  | 33 | 50 | mA |
| Supply current 2 | ICC2 | Stop mode |  | 3 | 5 | mA |
| Output saturation voltage | $\mathrm{V}_{\mathrm{O}}$ (sat) ${ }^{1}$ | $\mathrm{l}=1 \mathrm{~A}$ |  | 2.1 | 3.0 | V |
|  | $\mathrm{VO}\left(\right.$ sat) ${ }^{2}$ | $\mathrm{lO}=2 \mathrm{~A}$ |  | 3.0 | 4.2 | V |
| Output leak current | lo leak |  |  |  | 100 | $\mu \mathrm{A}$ |
| Voltage regulator |  |  |  |  |  |  |
| Output voltage | $\mathrm{V}_{\mathrm{H}}$ | $\mathrm{l} \mathrm{VH}=10 \mathrm{~mA}$ | 3.8 | 4.15 | 4.5 | V |
| Voltage variation | $\Delta \mathrm{V}_{\mathrm{H} 1}$ | $\mathrm{VCC}=9.5$ to 28 V |  | 60 | 150 | mV |
| Load variation | $\Delta \mathrm{V}_{\mathrm{H}} 2$ | IVH $=5$ to 20 mA |  | 60 | 150 | mV |
| Temperature coefficient |  |  |  | -2 |  | $\mathrm{mV} /{ }^{\circ} \mathrm{C}$ |
| Hall amp |  |  |  |  |  |  |
| Input bias current | IHB |  |  | 1 | 4 | $\mu \mathrm{A}$ |
| Common-mode input voltage | VICM |  | 1.5 |  | 2.8 | V |
| Hall input sensitivity |  |  | 100 |  |  | mVp-p |
| Hysteresis width | $\Delta \mathrm{V}_{\text {IN }}$ |  | 24 | 33 | 42 | mV |
| Low-to-high input voltage | VSLH |  | 8 | 20 | 32 | mV |
| High-to-low input voltage | VSHL |  | -25 | -13 | -1 | mV |
| Oscillator |  |  |  |  |  |  |
| High-level output voltage | V OH(CR) |  | 2.9 | 3.2 | 3.5 | V |
| Low-level output voltage | VOL(CR) |  | 0.9 | 1.1 | 1.3 | V |
| Oscillation amplitude |  |  | 1.8 | 2.1 | 2.4 | V |
| Oscillation frequency | f | $\mathrm{R}=30 \mathrm{k} \Omega, \mathrm{C}=1500 \mathrm{pF}$ |  | 18.5 |  | kHz |
| Temperature coefficient | $\Delta \mathrm{f}$ |  |  | 0.1 |  | \%/ ${ }^{\circ} \mathrm{C}$ |
| Comparator output voltage | Vosc | $\mathrm{losc}=20 \mathrm{~mA}$ |  |  | 1.5 | V |
| Current limiter |  |  |  |  |  |  |
| Limiter 1 | $\mathrm{V}_{\text {Rf }} 1$ |  | 0.42 | 0.5 | 0.6 | V |
| Limiter 2 | $\mathrm{V}_{\mathrm{Rf}} 2$ |  | 0.4 | 0.44 | 0.48 | V |
| Thermal shutdown |  |  |  |  |  |  |
| Thermal shutdown temperature | TSD | Design target, Note 1 | 150 | 180 |  | ${ }^{\circ} \mathrm{C}$ |
| Hysteresis width | $\Delta T S D$ |  |  | 30 |  | ${ }^{\circ} \mathrm{C}$ |
| Low-voltage protection voltage | VLVSD |  | 7.5 | 8.1 | 8.7 | V |
| Hysteresis width | $\Delta \mathrm{V}$ LVSD |  | 0.45 | 0.6 | 0.75 | V |
| FG amp |  |  |  |  |  |  |
| Input offset voltage | V IO(FG) |  | -10 |  | +10 | mV |
| Input bias current | IB (FG) |  | -1 |  | +1 | $\mu \mathrm{A}$ |
| High-level output voltage | $\mathrm{VOH}(\mathrm{FG})$ | $\mathrm{IFG}=-2 \mathrm{~mA}$ | 5.6 | 6.2 | 6.8 | V |
| Low-level output voltage | VoL(FG) | $\mathrm{I}_{\mathrm{FG}}=2 \mathrm{~mA}$ |  | 1 | 1.5 | V |
| FG input sensitivity |  | $10 \times$ Gain | 5 |  |  | mV |
| Schmitt width at next stage |  |  |  | 16 |  | mV |
| Operating frequency range |  |  |  |  | 5 | kHz |
| Open-loop voltage gain |  |  | 60 |  |  | dB |
| Speed discriminator |  |  |  |  |  |  |
| High-level output voltage | $\mathrm{VOH}(\mathrm{D})$ |  |  | 4.7 |  | V |
| Low-level output voltage | V OL(D) |  |  | 0.3 |  | V |
| Maximum clock frequency |  |  | 1.0 |  |  | MHz |
| Number of counts |  |  | 2044 | 2046 | 2048 |  |
| Integrator |  |  |  |  |  |  |
| Input offset voltage | VIO(INT) |  | -10 |  | +10 | mV |
| Input bias current | $\mathrm{I}_{\mathrm{B}(\mathrm{INT} \text { ) }}$ |  | -0.4 |  | +0.4 | $\mu \mathrm{A}$ |
| High-level output voltage | $\mathrm{VOH}(\mathrm{INT})$ |  | 3.7 | 4.3 | 4.9 | V |
| Low-level output voltage | VoL(INT) |  |  | 0.8 | 1.2 | V |
| Open-loop gain |  |  | 60 |  |  | dB |
| Gain-bandwidth product |  |  |  | 1.6 |  | MHz |
| Reference voltage |  |  | -5\% | V5/2 | 5\% | V |

Note 1: For parameters which have an entry of "design target value" in the "Conditions" column, no measurements are made.
Continued on next page.

## LB1820

Continued from preceding page.

| Parameter | Symbol | Conditions | min | typ | max | Unit |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 5 V supply | V5 |  | 4.6 | 5 | 5.4 | V |
| Lock detector |  |  |  |  |  |  |
| Low-level output voltage | V OL(LD) | $\mathrm{l}_{\mathrm{LD}}=10 \mathrm{~mA}$ |  |  | 0.5 | V |
| Lock range |  |  |  | $\pm 3.125$ |  | \% |
| Start/stop pin |  |  |  |  |  |  |
| Start/stop operating voltage |  |  | 0.4 | 0.5 | 0.6 | V |
| Crystal Oscillator |  |  |  |  |  |  |
| Precision of oscillating frequency |  | Referenced to indicated frequency | -500 |  | +500 | ppm |
| Temperature coefficient |  |  |  | -3 |  | ppm/ ${ }^{\circ} \mathrm{C}$ |
| Drift in rotational speed |  |  |  | $\pm 0.01$ |  | \% |

Truth Table

|  | Source $\rightarrow$ Sink | Input |  |  |
| :---: | :---: | :---: | :---: | :---: |
|  |  | IN1 | IN2 | IN3 |
| 1 | OUT3 $\rightarrow$ OUT2 | H | H | L |
| 2 | OUT3 $\rightarrow$ OUT1 | H | L | L |
| 3 | OUT2 $\rightarrow$ OUT1 | H | L | H |
| 4 | OUT2 $\rightarrow$ OUT3 | L | L | H |
| 5 | OUT1 $\rightarrow$ OUT3 | L | H | H |
| 6 | OUT1 $\rightarrow$ OUT2 | L | H | L |

## Pin Assignment




Internal Equivalent Circuit Block Diagram


## LB1820

## Pin Description

| Pin No. | Pin Name | Functions |
| :---: | :---: | :---: |
| 19, 20 | $\mathrm{IN}^{+1}, \mathrm{IN}^{-1}$ | OUT1: Hall element input pins for Phase 1. "H" logic is the state when $\mathrm{IN}^{+}>\mathrm{IN}^{-}$. |
| 17, 18 | $\mathrm{IN}^{+} 2, \mathrm{IN}^{-2}$ | OUT2: Hall element input pins for Phase 2. "H" logic is the state when $\mathrm{IN}^{+}>\mathrm{IN}^{-}$. |
| 15, 16 | $1 \mathrm{~N}^{+} 3, \mathrm{IN}^{-3}$ | OUT3: Hall element input pins for Phase 3. "H" logic is the state when $1 N^{+}>1 \mathrm{~N}^{-}$. |
| 6 | OUT 1 | Output pin 1. |
| 8 | OUT 2 | Output pin 2. |
| 10 | OUT 3 | Output pin 3. |
| 2 | $V_{C C}$ | Power supply for other than output blocks. |
| 12 | $\mathrm{V}_{\mathrm{M}}$ | Power supply for output blocks. |
| 11 | $\mathrm{R}_{\mathrm{f}}$ | Output current detection pin. $\mathrm{R}_{\mathrm{f}}$ is connected across this pin and GND to detect the output current as voltage. |
| 14 | GND | Ground for other than output blocks. <br> The lowest potential of output transistor is the voltage at $R_{f}$ pin. |
| 3 | CR | Sets the oscillating frequency of the switching regulator. |
| 1 | OSC | Outputs duty-controlled pulses. Open-collector output. |
| 24 | INTOUT | Integrator output pin (speed control pin). Varies the switching regulator output voltage. |
| 25 | INTIN | Integrator input pin. |
| 23 | Dout | Speed discriminator output pin. <br> Goes LOW when the specified speed is exceeded. |
| 4 | C | Suppresses ripples in the motor current during operation of current limiter 2. |
| 22 | LD | Lock detection pin. <br> Goes HIGH when the motor rotation speed is within the locking range. |
| 27 | FGIN ${ }^{-}$ | FG pulse input (Start/Stop control) pin. |
| 26 | FGIN ${ }^{+}$ | FG pulse input (4 V supply) pin. |
| 28 | FGout | FG amp output pin. |
| 21 | Xtal | Crystal oscillator connecting pin. |
| 13 | 5 V | 5 V supply pin. |

## Operation Notes

## Speed Control Circuit

This IC uses a speed discrimination circuit to perform speed control. The rotation accuracy of the speed discrimination method depends on the counter count. The counter count in this IC is 2046. On the FG1 cycle, a speed error signal with a resolution of $1 / 2046$ is output from the Dout pin (charge pump method).
The Dout output shifts among three states: high, high impedance, and low:

| High | : Output $S$ (acceleration signal) |
| :--- | :--- |
| High impedance | : When neither output $S$ nor output $F$ is output |
| Low | : Output $F$ (deceleration signal) |

The relationship between the FG frequency ( $\mathrm{f}_{\mathrm{FG}}$ ) and the quartz oscillation frequency ( fOSC ) can be calculated as follows:

$$
\begin{aligned}
\mathrm{f}_{\mathrm{FG}}= & \mathrm{fOSC} \div(\text { ECL division ratio } \times \text { count }) \\
& f^{\text {fOSC }} \div(8 \times 2046) \\
& f_{\mathrm{OSC}} \div 16368
\end{aligned}
$$

## PAM Drive System

This IC controls motor rotations by configuring an external switching regulator, and controlling the voltage $\left(\mathrm{V}_{\mathrm{M}}\right)$ of the regulator.
Select a switching regulator diode with a short reverse recovery time such as an FRD (First Recovery Diode). Because even a smooth coil can become a noise source, attention must be paid to the arrangement of components on the board (especially avoiding the effects of FG signal lines and integrated amplifiers).
Select a normal rectifier diode for the upper and lower motor drive pin section (OUT1 to 3).

## Current Limiter Circuit

The current limiter circuit consists of two limiter circuits.
(1) Limiter 1

Detection voltage $\mathrm{V}_{\mathrm{Rf}} 1=0.5 \mathrm{~V}$ typ. Current is limited by putting the lower output transistor in the nonsaturated state and then dropping the voltage applied to the motor.
(2) Limiter 2

Detection voltage $\mathrm{VRf}_{\mathrm{R}} 2=0.44 \mathrm{~V}$ typ. The $\mathrm{V}_{\mathrm{M}}$ voltage is limited by limiting the OSC pin "on duty" ratio.
Normally, if an excessive load is put on the motor, limiter 1 operates first, and after a delay in the switching regulator, limiter 2 operates.
Sometimes, after startup, the ASO of the output transistor is very severe. In such a case, it is necessary to perform a soft start (in which $\mathrm{V}_{\mathrm{M}}$ is increased gradually). When using soft starts, connect a capacitor between the pin ( $\mathrm{V}_{\mathrm{M}}, 5 \mathrm{~V}$, etc.) on which the voltage is to be increased during startup and the C pin. If soft starts are not to be used, connect a capacitor between the C pin and ground.

## Speed Lock Range

The speed lock signal is output from the LD pin. The speed lock range is within $\pm 3.13 \%$; if the motor rotations fall within the lock range, the LD pin goes low (open collector output).

## Start/stop Operation

The $\mathrm{FG}_{\mathrm{IN}}{ }^{-}$pin also serves as the start/stop pin. When the $\mathrm{FG}_{\mathrm{IN}}{ }^{-}$pin is connected to a transistor, etc., and the voltage is 0.5 V typ or less, the stop state goes into effect. In the stopped state, in addition to the drive outputs being turned off, the $\mathrm{FG}_{\mathrm{IN}}{ }^{+}, 5 \mathrm{~V}$, and other regulator outputs are also turned off.
When it is necessary to drive the motor at high speed, improvement is possible by adding a resistor (of approximately
$1 \mathrm{M} \Omega$ ) between FG out and $\mathrm{V}_{\mathrm{CC}}$. (The time from when the transistor is turned off until $\mathrm{FG}_{\mathrm{IN}}{ }^{-}$goes to 0.5 V is reduced.)

## Initial Reset Operation

At startup, it is possible to apply an initial reset to the logic circuits by delaying the increase in voltage on $\mathrm{FG}_{\mathrm{IN}}{ }^{-}$. If an initial reset is not applied, the LD pin may go low from start until the FG pulse is input to the logic circuits (until output of approximately $16 \mathrm{mVp}-\mathrm{p}$ is generated on FGout).
When an FG reset is applied, the capacitor between the $\mathrm{FG}_{\mathrm{IN}}{ }^{+}$and GND should be $4.7 \mu \mathrm{~F}$ or more (in order to delay the rise in $\mathrm{FG}_{\mathrm{IN}}{ }^{-}$). Caution is required, because if the FG amplifier input capacitor is too small and the feedback capacitor is too large, the reset time will be shorter. At start, a delay of about $5 \mu \mathrm{~s}$ or more from the rising edge of the 5 V regulator output until the $\mathrm{FG}_{\mathrm{IN}}{ }^{-}$voltage goes to 1.2 V is desirable.

## LB1820

## PWM Frequency Setting

The PWM frequency is determined by the resistor and capacitor connected to the CR pin. When a resistor is connected to the $\mathrm{FG}_{\text {IN }}{ }^{+}$pin, the PWM frequency can be roughly calculated by the following formula:
$\mathrm{fPWM} \approx 1 \div(1.2 \times \mathrm{C} \times \mathrm{R})$
The resistor must not be less than $30 \mathrm{k} \Omega$. It is desirable for the PWM frequency to be about 15 kHz .

## Quartz Oscillator

An oscillator, capacitor and resistor are to be connected to the quartz oscillator. When selecting the oscillator and the external capacitor and resistor, always obtain approval from the manufacturer of the oscillator in order to avoid problems.
(Circuit with external quartz oscillator)


External constants (reference values)

| Xtal (MHz) | $\mathrm{C} 1(\mathrm{pF})$ | $\mathrm{C} 2(\mathrm{pF})$ | $\mathrm{R}(\mathrm{k} \Omega)$ |
| :---: | :---: | :---: | :---: |
| 3 to 4 | 39 | 82 | 0.82 |
| 4 to 5 | 39 | 82 | 1.0 |
| 5 to 7 | 39 | 47 | 1.5 |
| 7 to 10 | 39 | 27 | 2.0 |

However, use a crystal such that the base wave fo impedance : 3fo impedance $=1: 5$ or more

When inputting external signals (of several MHz ) to the quartz oscillator, connect external components as shown in the diagram below.
$\mathrm{f}_{\mathrm{IN}}=1$ to 8 MHz
Input signal level
High level voltage: 4.0 V min.
Low level voltage: 1.5 V max.
$\mathrm{Ra}=2 \mathrm{k} \Omega, \mathrm{Rb}=1 \mathrm{k} \Omega$ (reference values)


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